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Signature design for oversaturated synchronous CDMA

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Direct Spreading CDMA

- Data stream of the k -th user $b_k(i)$, $i = \dots, -1, 0, 1, \dots$ modulates user-specific *signature* $\mathbf{s}_k = [s_k(0), s_k(1), \dots, s_k(N-1)]$
- All data streams and signatures are strictly time-aligned
- K - number of users
- $N = WT$ - spreading factor



Transmitted group signal

- $\mathbf{S} = [\mathbf{s}_1^T, \mathbf{s}_2^T, \dots, \mathbf{s}_K^T]$ – signature matrix
- A_k – amplitude of the k -th user
- $\mathbf{A} = \text{diag}[A_1, A_2, \dots, A_K]$ – diagonal amplitude matrix
- $\mathbf{b} = [b_1, b_2, \dots, b_K]$ – vector of current data bits
- Transmitted group signal:

$$\mathbf{y} = \sum_{k=1}^K A_k b_k \mathbf{s}_k = \mathbf{S}^T \mathbf{A} \mathbf{b}$$



Oversaturation

- Maximal number of orthogonal signatures is limited by $N = WT$
- Oversaturation (overloading) occurs when $K > N$
- Oversaturation efficiency is defined as $e_{ov} = K / N$



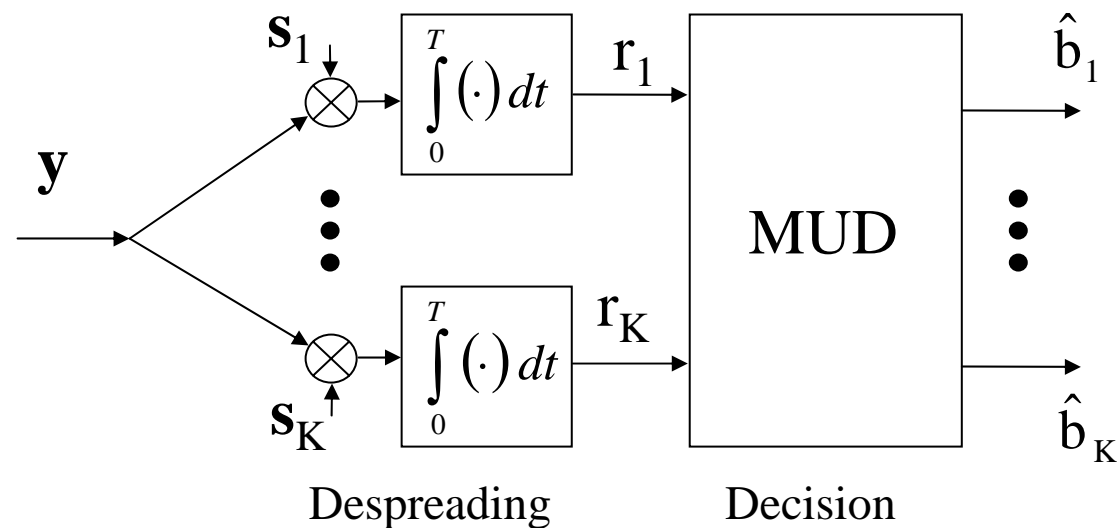
Oversaturation (cont.)

- With oversaturation, orthogonal signaling is not possible \Rightarrow unavoidable MAI occurs
- Conventional receiver is not applicable \Rightarrow multiuser receiver must be used
- The problem of optimal multiuser is its exponential complexity in the number of users (if no special care is taken in signature design)



Optimal multiuser receiver (cont.)

- Bank of matched filters followed by multiuser detector (MUD). Matched filter outputs are $\mathbf{r} = \mathbf{S}(\mathbf{S}^T \mathbf{A} \mathbf{b} + \mathbf{n}) = \mathbf{R} \mathbf{A} \mathbf{b} + \mathbf{n}'$





Optimal multiuser receiver (cont.)

- Optimal multiuser detector calculates decision

$$\hat{\mathbf{b}} = \arg \max \left(2\mathbf{r}^T \mathbf{A}\mathbf{b} - \mathbf{b}^T \mathbf{A}\mathbf{R}\mathbf{A}\mathbf{b} \right)$$

- Simplifications are possible, if e.g. signatures have hierarchical cross-correlation matrix structure, Learned's algorithm having polynomial complexity can be applied.
- Optimal signature set provides maximal separation of all possible group signals



Group Orthogonal CDMA

- Another approach to simplify receiver: divide users to groups (subspaces) that don't interfere and oversaturate groups => MAI only inside groups
- Formally, signal space dimension N is divided to subspaces having dimension L . Each subspace is oversaturated by s users. The number of users served $K=N(L+s)/L=N+N_s/L$, $e_{ov}=1+s/L$
- Abbreviation: GO-CDMA (group orthogonal code division multiple access)
- It is possible to use optimal multiuser detector for each group due to small number of users inside group



GO-CDMA signatures

- Signature design performed for $L=2,3,4$ and $s=1$
 - $L>4$ cannot improve maximum minimum Euclidean distance further
 - $s>1$ did not produce good results
 - Symmetrical group signal set
 - $L=4, s=1$ results in R-T signatures



GO-CDMA signatures

- H matrix with orthogonal rows $\mathbf{H}_N = \begin{pmatrix} \mathbf{h}_1 \\ \vdots \\ \mathbf{h}_N \end{pmatrix}$
- L=2, i-th group:

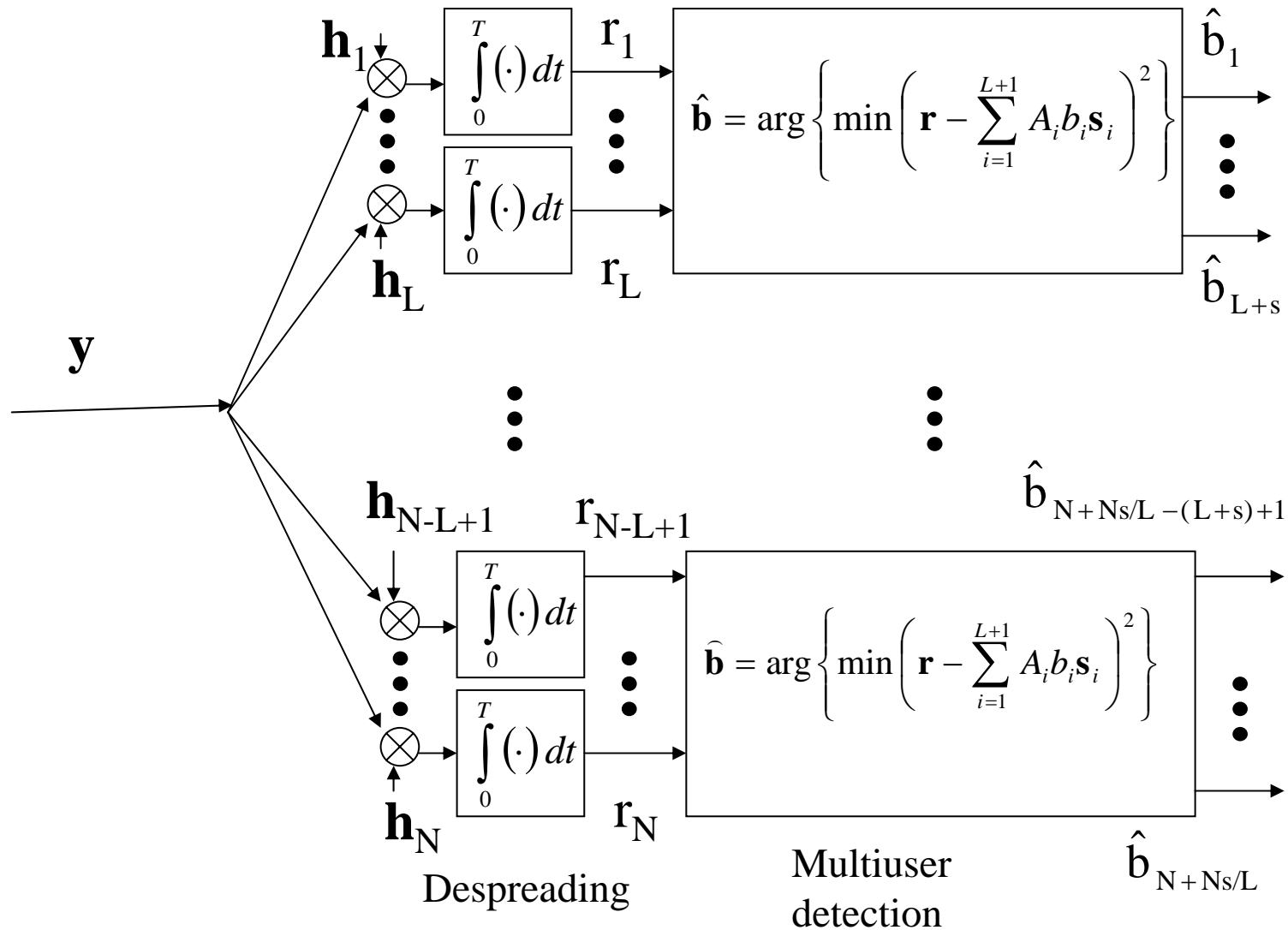
$$\mathbf{s}_i = \begin{pmatrix} s_1^i \\ s_2^i \\ s_3^i \end{pmatrix} = \frac{1}{\sqrt{2(1+\alpha^2)}} \begin{pmatrix} \sqrt{2} & \alpha\sqrt{2} \\ \alpha\sqrt{2} & \sqrt{2} \\ \sqrt{1+\alpha^2} & \sqrt{1+\alpha^2} \end{pmatrix} \begin{pmatrix} \mathbf{h}_{2i-1} \\ \mathbf{h}_{2i} \end{pmatrix}, \alpha = \frac{4 - \sqrt{10 + 2\sqrt{5}}}{\sqrt{5} - 1}$$

- L=3, i-th group:

$$\mathbf{s}_i = \begin{pmatrix} s_1^i \\ s_2^i \\ s_3^i \\ s_4^i \end{pmatrix} = \frac{1}{\sqrt{3(1+2\beta^2)}} \begin{pmatrix} \sqrt{3} & \beta\sqrt{3} & \beta\sqrt{3} \\ \beta\sqrt{3} & \sqrt{3} & \beta\sqrt{3} \\ \beta\sqrt{3} & \beta\sqrt{3} & \sqrt{3} \\ \sqrt{1+2\beta^2} & \sqrt{1+2\beta^2} & \sqrt{1+2\beta^2} \end{pmatrix} \begin{pmatrix} \mathbf{h}_{3i-2} \\ \mathbf{h}_{3i-1} \\ \mathbf{h}_{3i} \end{pmatrix}, \beta = \frac{-12 + (1 + \sqrt{7})\sqrt{14 - \sqrt{7}}}{16 - 2\sqrt{7}}$$



GO-CDMA receiver(s)



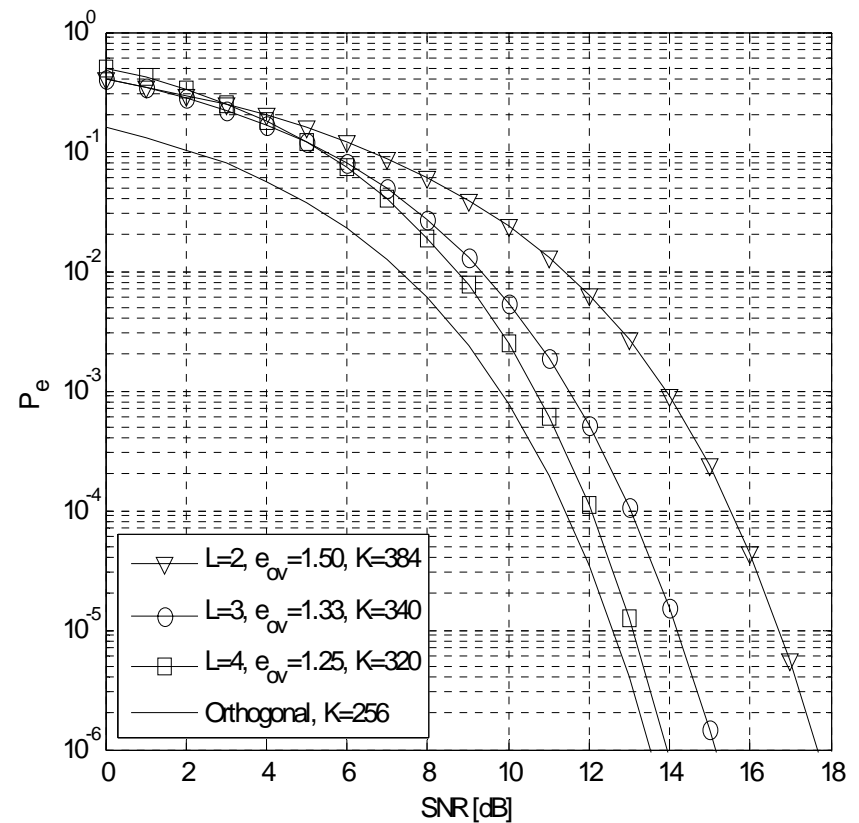


Performance of GO-CDMA

Euclidean distance

| L | d^2_{\min} | Loss [dB] | e_{ov} |
|---|--------------|-----------|----------|
| 1 | 4.00 | 0 | 1.00 |
| 2 | 1.53 | 4.18 | 1.50 |
| 3 | 2.71 | 1.69 | 1.33 |
| 4 | 4.00 | 0 | 1.25 |

Bit error rate (BER)





Collaboratively coded GO- CDMA (CCGO-CDMA)

- S-CDMA can be used in downlink of cellular system or in uplink in application of channelization (separation of channels of one user)
 - Transmitter controls all transmissions
 - Requiring signature per user seems unnecessary restriction
 - It is possible to collaboratively encode the data of users inside one group
 - Subspace acts as a common signature for users inside group



Signatures for CCGO-CDMA

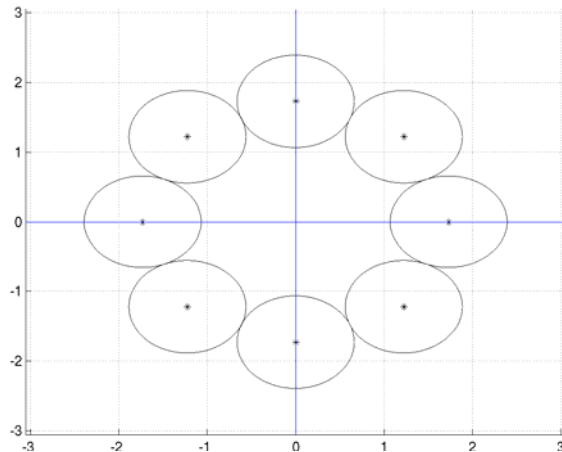
- Data bits of $L+s$ users are jointly mapped into 2^{L+s} L -dimensional signal vectors
- Signature design performed for $L=2,3,4,5$ and $s=1,2$. Now, it is possible to improve performance with $L>4$ and $s>1$.
- Sphere packing theory assists in finding good signal constellations. Signals treated as L -dimensional spheres having radius $d_{\min} / 2$.
- Few different options investigated:
 - maximal minimum distance constellation
 - symmetrical constellations with or without zero-vector
 - equal energy constellation
- Modulation mapping not a trivial problem



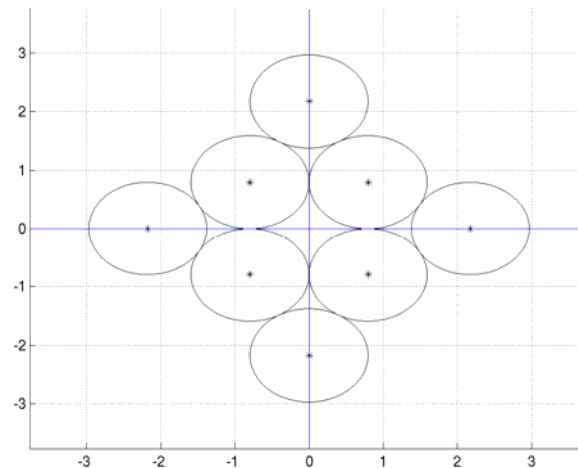
CCGO-CDMA Signatures

$d_{\min}^2 = 1.76 \Rightarrow \text{loss} = 3.76 \text{ dB}$ **($L=2, s=1$)** $d_{\min}^2 = 2.54 \Rightarrow \text{loss} = 1.98 \text{ dB}$

A.

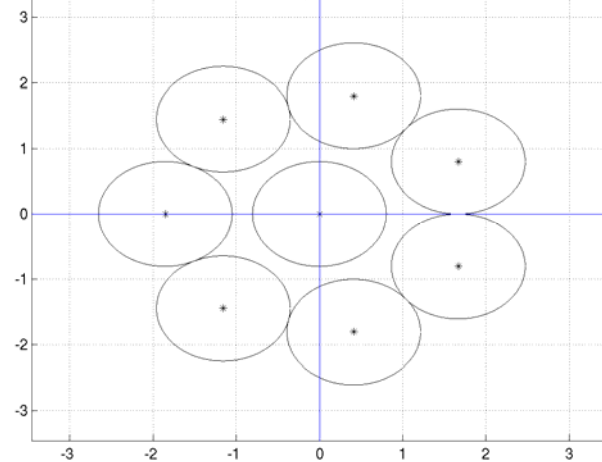


B.



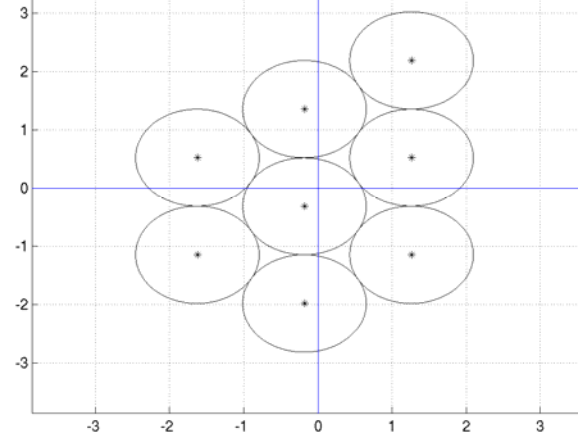
$d_{\min}^2 = 2.58 \Rightarrow \text{loss} = 1.90 \text{ dB}$

C.



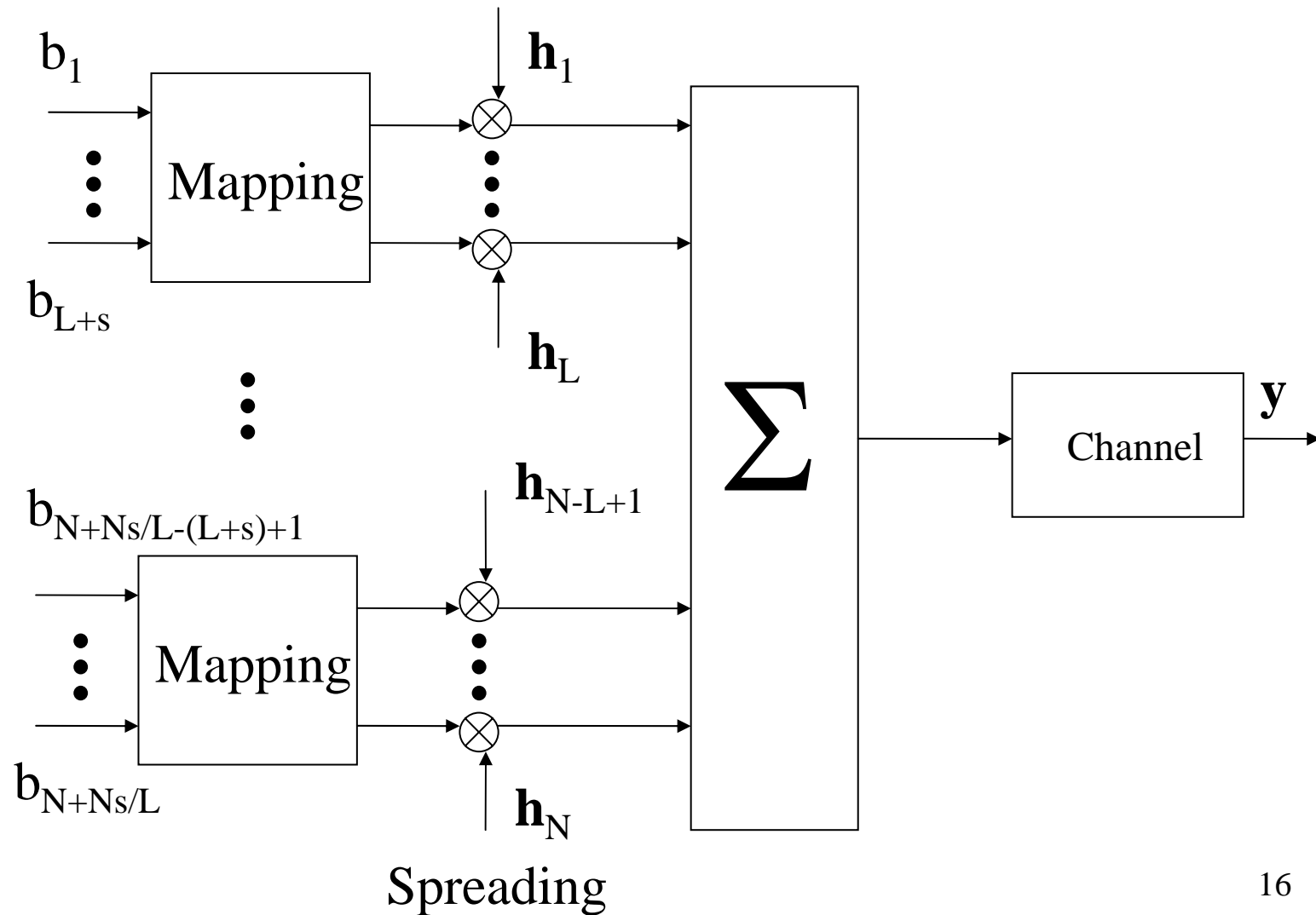
D.

$d_{\min}^2 = 2.78 \Rightarrow \text{loss} = 1.58 \text{ dB}$



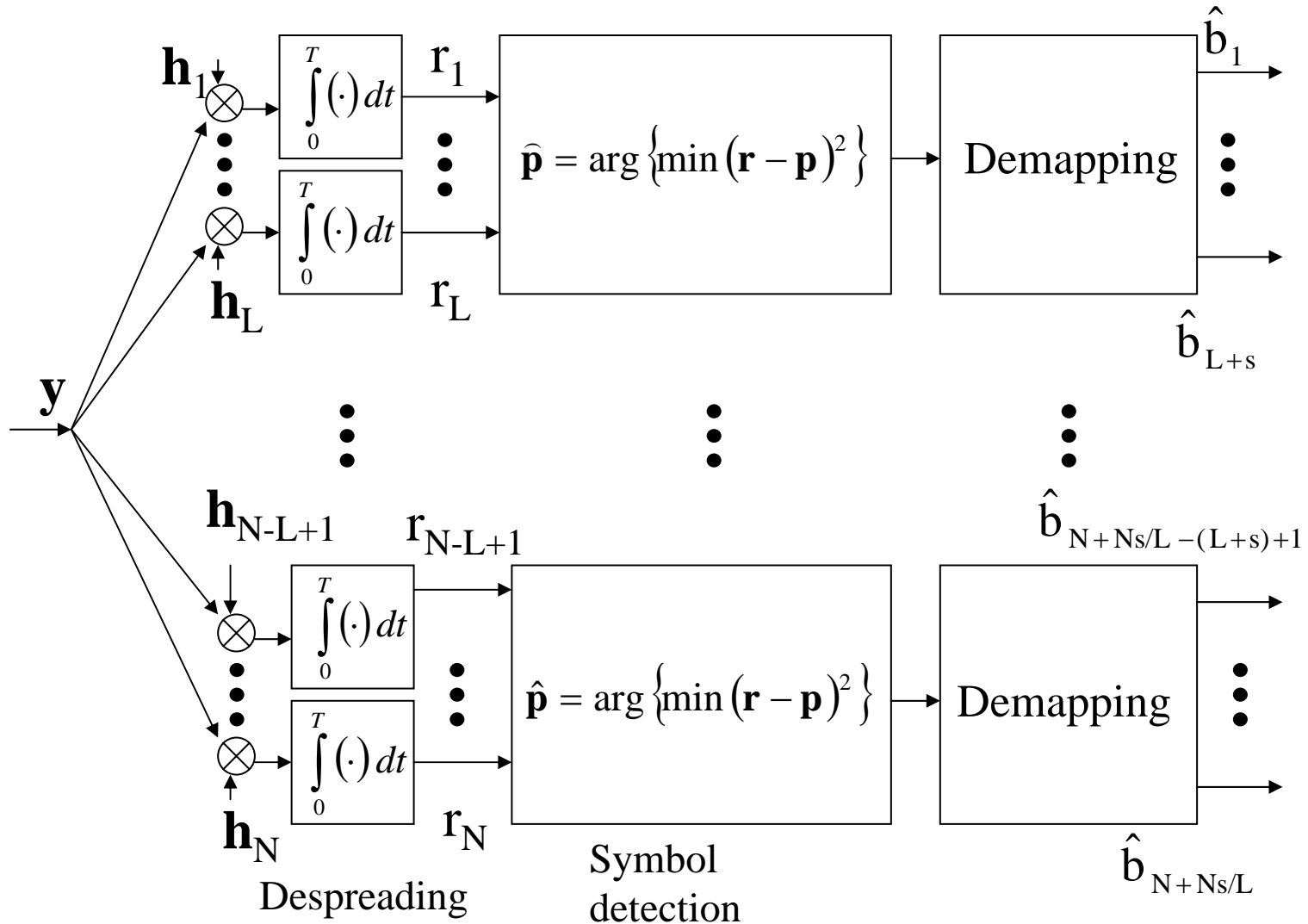


CCGO-CDMA transmitter





CCGO-CDMA Receiver(s)





Performance of CCGO-CDMA (Euclidean distance)

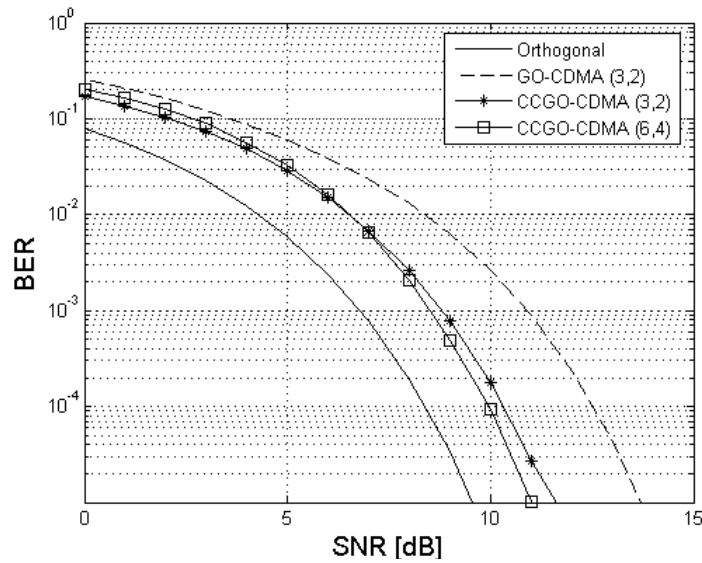
| Const. | L=2, s=1 | | L=3, s=1 | | L=4, s=1 | | L=5, s=1 | |
|--------|--------------|------|--------------|------|--------------|-------|--------------|-------|
| | d^2_{\min} | loss | d^2_{\min} | loss | d^2_{\min} | loss | d^2_{\min} | loss |
| A | 1.76 | 3.57 | 3.10 | 1.10 | 4.04 | -0.05 | 4.73 | -0.73 |
| B | 2.54 | 1.98 | 3.25 | 0.90 | 4 | 0 | 4.36 | -0.38 |
| C | 2.58 | 1.90 | 3.45 | 0.61 | 4.22 | -0.24 | 4.83 | -0.82 |
| D | 2.78 | 1.58 | 3.63 | 0.42 | 4.22 | -0.23 | 4.47 | -0.48 |

| Const. | L=4, s=2 | | L=5, s=2 | |
|--------|--------------|------|--------------|-------|
| | d^2_{\min} | loss | d^2_{\min} | loss |
| A | 3.13 | 1.07 | 3.92 | 0.08 |
| B | 3.20 | 0.97 | 4.15 | -0.16 |
| C | 3.19 | 0.99 | 3.97 | 0.03 |
| D | 3.28 | 0.86 | 4.19 | -0.20 |

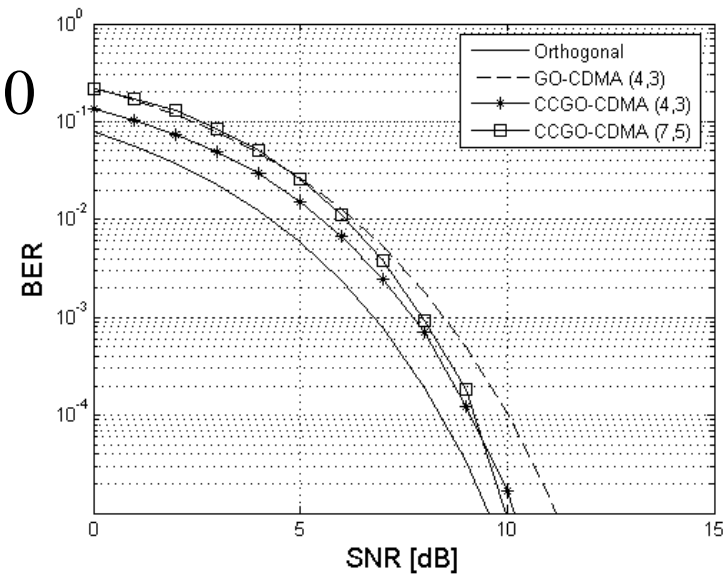
- Negative loss implies gain
- The order of superiority changes when e_{ov} decreases



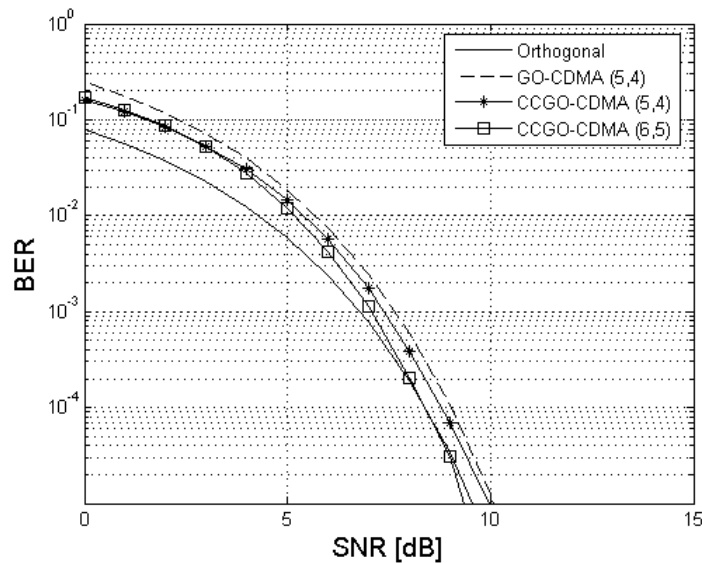
BER of constellation type C



$e_{ov}=1.50$



$e_{ov}=1.33$ and 1.40



$e_{ov}=1.25$ and 1.20